

Write in each answer paper your name, department, student number, the course name and code, and the date. Please tell if you want to pass the course as a postgraduate version. Number each paper you submit and denote the total no. of pages. 5 problems, 30 points total. Papers in English only. Please feel free to answer in English, Finnish or Swedish.

1. (1p each) Define and describe *briefly* (2..3 lines of text) the following concepts:
  - a) Synchronous sampling
  - b) GPS
  - c) NDA phase estimation
  - d) S-curve
  - e) Fractional delay filter
  - f) VCO
  
2. (6p) Assume that we wish to estimate the amplitude  $A$  of a sinusoid embedded in white Gaussian noise (WGN)  $w(k)$ :

$$r(k) = A \cos(2\pi f_0 k + \theta) + w(k), \quad k = 0, 1, \dots, N-1 \quad (1)$$

where  $r(k)$  is the discrete-time received signal sequence of  $N$  samples. The frequency  $f_0$  and phase  $\theta$  are assumed known. The probability density function (PDF) is

$$p(\mathbf{r}; A) = \frac{1}{(2\pi\sigma^2)^{\frac{N}{2}}} \exp\left\{-\frac{1}{2\sigma^2} \sum_{k=0}^{N-1} [r[k] - A \cos(2\pi f_0 k + \theta)]^2\right\} \quad (2)$$

where  $\sigma^2$  is the noise variance. Find the Cramer-Rao bound for the estimator for  $A$  in closed form. (Hint: the Cramer-Rao bound for a scalar parameter is defined as the negative inverse of the second derivative of the PDF with respect to the parameter in question.)

3. Consider the carrier recovery problem in a 4-PSK system. Assume that after the first (analog) carrier demodulation stage a low-frequency modulating sinusoid still remains so that the appropriate baseband signal model is

$$r(t) = e^{j(2\pi f t + \theta)} \sum_i a_i h_T(t - iT - \tau) + w(t) \quad (3)$$

where  $f$  is the carrier frequency (error) to be estimated, carrier phase  $\theta$  is unknown but the timing error  $\tau$  is known. Data symbols  $a_k$  are known because a training signal is used at the beginning of transmission.

- (a) (3p) Derive the matched filter output  $z(k) = a_k^* y(k) = e^{j[2\pi f(kT + \tau) + \theta]} + n'(k)$  that can be developed under certain assumptions (Nyquist pulse, small enough  $f$ ), as explained in the lecture (slides).
- (b) (3p) You have two samples of  $z(k)$  available. Develop an estimator for the carrier frequency.

4. (9p TOTAL - SPECIAL BONUS PROBLEM!) An  $N$ th-order allpass filter has a transfer function of the form

$$H(z) = \frac{z^{-N} A(z^{-1})}{A(z)} = \frac{a_N + \dots + a_1 z^{-N+1} + z^{-N}}{1 + a_1 z^{-1} + \dots + a_N z^{-N}} \quad (4)$$

a) (4p) Design a *first-order allpass interpolator* such that its group delay equals to  $(1+\mu)$  at the zero frequency. Check also that your filter is stable!

Hint: the group delay of the allpass filter is determined by the denominator group delay so that

$$\begin{aligned} \tau_{g,H}(\omega) &= N + 2\tau_{g,A}(\omega) \\ \tau_{g,A}(\omega) &= \frac{-d\theta_A(\omega)}{d\omega}, \quad \theta_A(\omega) = \arg\{A(e^{j\omega})\} \end{aligned} \quad (5)$$

(Note: the signs may be defined slightly differently from what you have seen before – be careful!)

b) (3p) Determine a 2-tap Lagrange interpolator such that its group delay equals to  $\mu$  at the zero frequency.

c) (2p) Compare your allpass and Lagrange interpolators and discuss their properties and suitability for symbol timing applications.

5. (6p) Channel estimation (CE) principles: explain the difference between channel equalization and CE. What is blind CE?