

Write in each answer paper your name, department, student number, the course name and code, and the date. Number each paper you submit and denote the total no. of pages. 5 problems, 30 points total. Papers in English only. The BETA mathematical tables can be utilized – you can borrow a copy from the exam supervisor if you do not have your own. A basic calculator can be used (no memory, no graphics).

The homework bonus will be valid for possible future exams too.

1. (1p each) Define and describe *briefly* (2..3 lines of text) the following concepts:

- a) Echo cancellation
- b) Matched filter
- c) Nyquist criterion
- d) Channel capacity
- e) OFDM
- f) ZF equalizer

2. Nonlinear equalizers (6p):

Describe and compare the DFE equalizer and the Viterbi algorithm (the latter provided with a simple channel estimator). Discuss the advantages and disadvantages applications like in mobile communications channels (with fading) and wireline channel.

3. Adaptive filters (6p):

Let us consider a discrete-time model for a communication system in a linear channel (sampled at the symbol rate). The received signal samples  $r(k)$  are filtered by an  $N$ -tap FIR filter (equalizer). The equalizer output can be expressed as

$$y(k) = \mathbf{h}_R^T \mathbf{r}(k). \quad (1)$$

where  $\mathbf{h}_R$  and  $\mathbf{r}$  are  $N$ -dimensional column vectors. The mean squared error (MSE) can be expressed as

$$E[e^2(k)] = E[a_k^2] - 2\mathbf{h}_R^T \mathbf{p}_0 + \mathbf{h}_R^T \mathbf{R} \mathbf{h}_R \quad (2)$$

where  $\mathbf{p} = E[\mathbf{r}(k)a_k]$  and  $\mathbf{R} = E[\mathbf{r}(k)\mathbf{r}^T(k)]$ .

- a) (2p) Derive the optimal minimum-MSE equalizer.
- b) (2p) The general structure for an adaptive equalizer using *MSE gradient* (MSEG) algorithm is of the form

$$\mathbf{h}_R[j+1] = \mathbf{h}_R[j] - \frac{\beta}{2} \nabla_{\mathbf{h}_R} E[e^2(k)]. \quad (3)$$

Derive the MSEG algorithm for the equalizer.

- c) (2p) Consider a 1-tap adaptive MSEG filter. Derive the criterion for the adaptation convergence in terms of the step parameter  $\beta$ .

**4. Nyquist and Matched filtering criteria (9p bonus):**

Let us derive pulse waveforms which meet the Nyquist criterion:

- a) (3p) Assume an ideal baseband communication system of data rate  $1/T$  in an AWGN channel which needs no excess bandwidth. Define the *pulse spectrum* and derive the *continuous-time pulse waveform* via inverse Fourier transform. No filtering assumed in the receiver, just symbol-rate sampling.
- b) (3p) The same as a) except now we assume excess bandwidth  $\alpha$  ( $0 < \alpha < 1$ ) and that the spectrum is *piecewise constant*.
- c) (3p) The same as b) except now we want to design transmit and receive filters that form a matched-filter pair and whose *convolution meets the Nyquist criterion*. Assume again that the *spectrum of the convolution is piecewise constant*.

**5. Channel capacity (6p):**

Consider the transmission of signal  $x(t)$  over a linear channel with associated impulse response  $c(t)$  and frequency response  $C(f)$ . The output waveform of the channel is then  $r(t) = c(t)*x(t)$ , where '\*' denotes convolution. The output of the channel is thereafter corrupted by colored noise  $n(t)$  with power spectral density (PSD)  $S_n(f)$ .

- a) (4p) Solve for the optimum transmit power spectrum  $S_x(f)$  that maximizes the channel capacity when the total transmit power  $P_x$  is limited to 12 W (see Equation (5) on Page 3). To help you solve the problem, Figure 1 below provides you with the PSD of the noise  $S_n(f)$  and the magnitude squared of the channel transfer function. Assume that  $S_n(f) = \infty$ , for  $|f| > 4$  Hz.
- b) (2p) Determine the channel capacity resulting from the optimized transmit power spectrum in a).

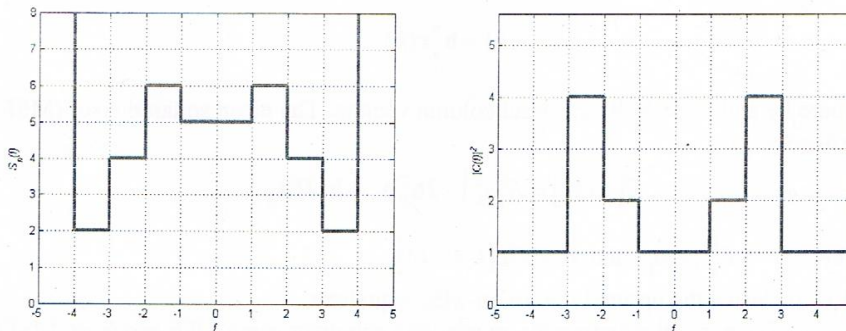


Figure 1: PSD of the noise  $S_n(f)$  in W/Hz and channel gain  $|C(f)|^2$  versus frequency  $f$  in Hz.

Hint: The optimal power spectrum is obtained with the water-pouring theorem as:

$$S_{x,opt}(f) = L - S_n(f) / |C(f)|^2 \quad (4)$$

whenever resulting  $S_{x,opt}(f)$  is positive (zero otherwise) and the water-filling  $L$  is determined so that the total transmit power is limited, i.e.,

$$P_x = \int_{-\infty}^{\infty} S_{x,opt}(f) df \quad (5)$$

The optimal capacity is then obtained by (double-sided) integration:

$$C_{CH} = \frac{1}{2} \int_{-\infty}^{\infty} \log_2 \left( 1 + \frac{S_{x,opt}(f) |C(f)|^2}{S_n(f)} \right) df \quad (6)$$